CONTROL REGISTER

The ACIA Control Register consists of eight bits of write-only buffer that are selected when RS and R/$\overline{W}$ are low. This register controls the function of the receiver, transmitter, interrupt enables, and the Request-to-Send peripheral/modem control output.

Counter Divide Select Bits (CR0 and CR1) — The Counter Divide Select Bits (CR0 and CR1) determine the divide ratios utilized in both the transmitter and receiver sections of the ACIA. Additionally, these bits are used to provide a master reset for the ACIA which clears the Status Register (except for external conditions on $\overline{CTS}$ and $\overline{DCD}$) and initializes both the receiver and transmitter. Master reset does not affect other Control Register bits. Note that after power-on or a power fail/restart, these bits must be set high to reset the ACIA. After resetting, the clock divide ratio may be selected. These counter select bits provide for the following clock divide ratios:

CR1	CR0	Function
0	0	÷1
0	1	÷16
1	0	÷64
1	1	Master Reset

Word Select Bits (CR2, CR3, and CR4) — The Word Select bits are used to select word length, parity, and the number of stop bits. The encoding format is as follows:

CR4	CR3	CR2	Function
0	0	0	7 Bits + Even Parity + 2 Stop Bits
0	0	1	7 Bits + Odd Parity + 2 Stop Bits
0	1	0	7 Bits + Even Parity + 1 Stop Bit
0	1	1	7 Bits + Odd Parity + 1 Stop Bit
1	0	0	8 Bits + 2 Stop Bits
1	0	1	8 Bits + 1 Stop Bit
1	1	0	8 Bits + Even parity + 1 Stop Bit
1	1	1	8 Bits + Odd Parity + 1 Stop Bit

Word length, Parity Select, and Stop Bit changes are not buffered and therefore become effective immediately.

Transmitter Control Bits (CR5 and CR6) — Two Transmitter Control bits provide for the control of the interrupt from the Transmit Data Register Empty condition, the Request-to-Send ($\overline{RTS}$) output, and the transmission of a Break level (space). The following encoding format is used:

CR6	CR5	Function
0	0	$\overline{RTS}$ = low, Transmitting Interrupt Disabled.
0	1	$\overline{RTS}$ = low, Transmitting Interrupt Enabled.
1	0	$\overline{RTS}$ = high, Transmitting Interrupt Disabled.
1	1	$\overline{RTS}$ = low, Transmits a Break level on the Transmit Data Output. Transmitting Interrupt Disabled.

Receive Interrupt Enable Bit (CR7) — The following interrupts will be enabled by a high level in bit position 7 of the Control Register (CR7): Receive Data Register Full, Overrun, or a low-to-high transition on the Data Carrier Detect ($\overline{DCD}$) signal line.

STATUS REGISTER

Information on the status of the ACIA is available to the MPU by reading the ACIA Status Register. This read-only register is selected when RS is low and R/$\overline{W}$ is high. Information stored in this register indicates the status of the Transmit Data Register, the Receive Data Register and error logic, and the peripheral/modem status inputs of the ACIA.

Receive Data Register Full (RDRF), Bit 0 — Receive Data Register Full indicates that received data has been transferred to the Receive Data Register. RDRF is cleared after an MPU read of the Receive Data Register or by a master reset. The cleared or empty state indicates that the contents of the Receive Data Register are not current. Data Carrier Detect being high also causes RDRF to indicate empty.

Transmit Data Register Empty (TDRE), Bit 1 — The Transmit Data Register Empty bit being set high indicates that the Transmit Data Register contents have been transferred and that new data may be entered. The low state indicates that the register is full and that transmission of a new character has not begun since the last write data command.

Data Carrier Detect (DCD), Bit 2 — The Data Carrier Detect bit will be high when the $\overline{DCD}$ input from a modem has gone high to indicate that a carrier is not present. This bit going high causes an Interrupt Request to be generated when the Receive Interrupt Enable is set. It remains high after the $\overline{DCD}$ input is returned low until cleared by first reading the Status Register and then the Data Register or until a master reset occurs. If the $\overline{DCD}$ input remains high after read status and read data or master reset has occurred, the interrupt is cleared, the DCD status bit remains high and will follow the $\overline{DCD}$ input.

Clear-to-Send ($\overline{CTS}$), Bit 3 — The Clear-to-Send bit indicates the state of the Clear-to-Send input from a modem. A low $\overline{CTS}$ indicates that there is a Clear-to-Send from the modem. In the high state, the Transmit Data Register Empty bit is inhibited and the Clear-to-Send status bit will be high. Master reset does not affect the Clear-to-Send status bit.

Framing Error (FE), Bit 4 — Framing error indicates that the received character is improperly framed by a start and a stop bit and is detected by the absence of the first stop bit. This error indicates a synchronization error, faulty transmission, or a break condition. The framing error flag is set or reset during the receive data transfer time. Therefore, this error indicator is present throughout the time that the associated character is available.

Receiver Overrun (OVRN), Bit 5 — Overrun is an error flag that indicates that one or more characters in the data stream were lost. That is, a character or a number of characters were received but not read from the Receive Data Register (RDR) prior to subsequent characters being received. The overrun condition begins at the midpoint of the last bit of the second character received in succession without a read of the RDR having occurred. The Overrun does not occur in the Status Register until the valid character prior to Overrun has

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been read. The RDRF bit remains set until the Overrun is reset. Character synchronization is maintained during the Overrun condition. The Overrun indication is reset after the reading of data from the Receive Data Register or by a Master Reset.

Parity Error (PE), Bit 6 — The parity error flag indicates that the number of highs (ones) in the character does not agree with the preselected odd or even parity. Odd parity is defined to be when the total number of ones is odd. The parity error indication will be present as long as the data character is in the RDR. If no parity is selected, then both the transmitter parity generator output and the receiver partiy check results are inhibited.

Interrupt Request (IRQ), Bit 7 — The IRQ bit indicates the state of the $\overline{IRQ}$ output. Any interrupt condition with its applicable enable will be indicated in this status bit. Anytime the $\overline{IRQ}$ output is low the IRQ bit will be high to indicate the interrupt or service request status. $\overline{IRQ}$ is cleared by a read operation to the Receive Data Register or a write operation to the Transmit Data Register.

ORDERING INFORMATION

Package Type	Frequency (MHz)	Temperature	Order Number
Cerdip	1.0	0°C to 70°C	MC6850S
S Suffix	1.0	−40°C to 85°C	MC6850CS
	1.5	0°C to 70°C	MC68A50S
	1.5	−40°C to 85°C	MC68A50CS
	2.0	0°C to 70°C	MC68B50S
Plastic	1.0	0°C to 70°C	MC6850P
P Suffix	1.0	−40°C to 85°C	MC6850CP
	1.5	0°C to 70°C	MC68A50P
	1.5	−40°C to 85°C	MC68A50CP
	2.0	0°C to 70°C	MC68B50P

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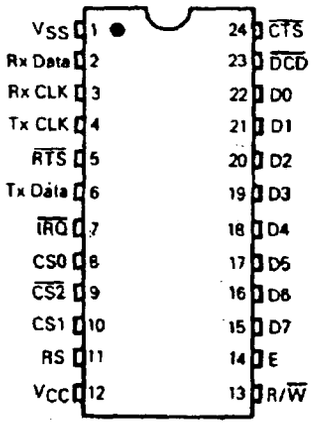
PIN ASSIGNMENTS



IM4702/4712
Baud Rate Generator

GENERAL DESCRIPTION

The IM4702/12 Baud Rate Generators provide necessary clock signals for digital data transmission systems, such as UARTs, using a 2.4576MHz crystal oscillator as an input. They control up to 8 output channels and can be cascaded for output expansion.

Output rate is controlled by four digital input lines, and with the specified crystal, is selectable from "zero" through 9600 Baud. In addition, 19200 Baud is possible via hardwiring.

Multi-channel operation is facilitated by making the clock frequency and the ÷8 prescaler outputs available externally. This allows up to eight simultaneous Baud rates to be generated.

The IM4712 is identical to the IM4702 with the exception that the IM4712 integrates the oscillator feedback resistor and two load capacitors on-chip.

FEATURES

- **Provides 14 Most Commonly Used BAUD Rates**
- **On-Chip Oscillator Requires Only One External Part (IM4712)**
- **Controls Up to Eight Transmission Channels**
- **TTL Compatible Outputs Will Sink 1.6mA**
- **Uses Standard 2.4576MHz Crystal**
- **Low Power Consumption: 5.5mW Guaranteed Maximum Standby**
- **Pin and Function Compatible With 4702B and HD-4702**
- **Inputs Feature Active Pull-Ups**

ORDERING INFORMATION

Order Number	Temperature Range	Package
IM4702IJE	−40°C to +85°C	16-pin CERDIP
IM4702IPE	−40°C to +85°C	16-pin PLASTIC
IM4712IJE	−40°C to +85°C	16-pin CERDIP
IM4712IPE	−40°C to +85°C	16-pin PLASTIC

PIN DESCRIPTION

Signal	Pin	Description
$Q_0 - Q_2$	1,2,3	Prescaler Outputs
$\overline{ECP}$	4	External Clock Enable Input
CP	5	External Clock Input
O_X	6	Crystal Output
I_X	7	Crystal Input
V_{SS}	8	Negative Supply
C_0	9	Clock Output
Z	10	Baud Rate Output
$S_0 - S_3$	14-11	Baud Rate Select Inputs
I_M	15	Multiplexed Input
V_{DD}	16	Positive Supply

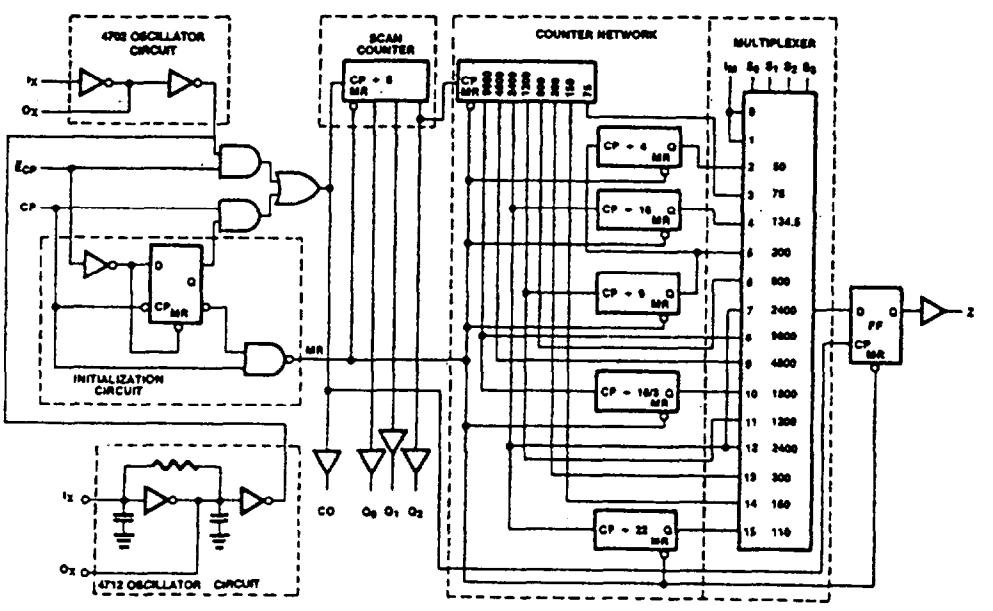


Figure 1: Functional Diagram

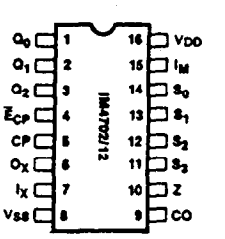


Figure 2: Pin Configuration (Outline drawings JE, PE)

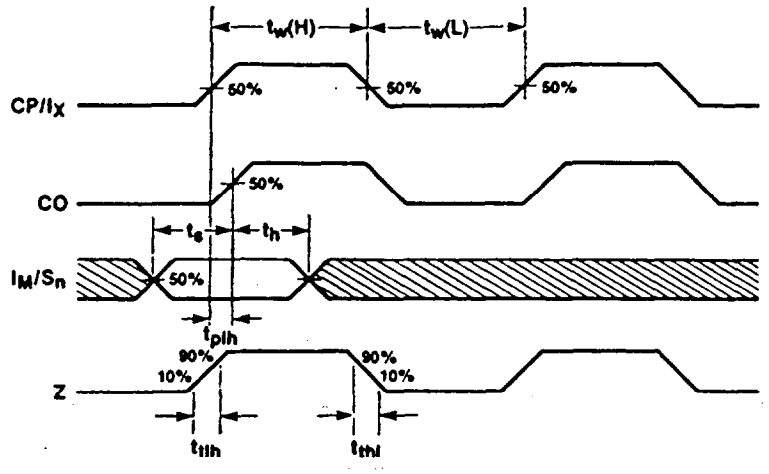


Figure 3: Switching Waveforms

FUNCTIONAL DESCRIPTION

Digital data transmission systems employ a wide range of standardized bit rates, ranging from 50 baud (for electromechanical devices) to 9600 baud (for high speed modems). Modern electronic systems commonly use Universal Asynchronous Receiver and Transmitter circuits (UARTs) to convert parallel data inputs into a serial bit stream (transmitter) and to reconvert the serial bit stream into parallel outputs (receiver). In order to resynchronize the incoming serial data, the reciever requires a clock rate which is a multiple of the incoming bit rate. Popular MOSLSI UART circuits use a clock that is 16 times the transmitted bit rate. The IM4702/12 can generate 14 standard clock rates from one common high frequency input.

The IM4702/12 contains the following five function subsystems.

Oscillator — For conventional operation generating 16 output clock pulses per bit period, the input clock frequency must be 2.4576MHz (i.e. 9600 baud x 16 x 16, since the scan counter and the first flip-flop of the counter chain act as an internal ÷ 16 prescaler). A lower input frequency will result in a proportionally lower output frequency.

The IM4702/12 can be driven from two alternate clock sources: (1) When the $\overline{E}_{cp}$ (External Clock Enable) input is LOW, the CP input is the clock source. (2) When the $\overline{E}_{cp}$ input is HIGH, a crystal connected between I_x and O_x, or a signal applied to the I_x input, is the clock source.

Prescaler (Scan Counter) — The clock frequency is made available on the CO (Clock Output) pin and is applied to the ÷ 8 prescaler with buffered outputs Q_0, Q_1, and Q_2.

Table 1: Clock Modes and Initialization

I_x	$\overline{E}_{CP}$	CP	Operation
⎍⎍⎍	H	L	Clocked from I_x
X	L	⎍⎍⎍	Clocked from CP
X	H	H	Continuous Reset
X	L	⎍	Reset During First CP = HIGH Time

H = HIGH Level
L = LOW Level
X = Don't Care
⎍ = 1st HIGH Level Clock Pulse After $\overline{E}_{CP}$ Goes LOW
⎍⎍⎍ = Clock Pulses

0373-8

Counter Network — The prescaler output Q_2 is a square wave of 1/8 the input frequency, and is used to drive the frequency counter network generating 13 standardized frequencies. Note that the frequencies are labeled in the block diagram and described in terms of the transmission bit rate. In a conventional system using a 2.4576MHz clock input, the actual output frequencies are 16 times higher.

The output from the first frequency divider flip-flop is thus labeled 9600, since it is used to transmit or receive 9600 baud (bits per second). The actual frequency at this node is 16 x 9.6kHz = 153.6kHz. Seven more cascaded binaries generate the appropriate frequencies for bit rates 4800, 2400, 1200, 600, 300, 150, and 75.

The other five bit rates are generated by individual counters:

bit rate 1200 is divided by 6 to generate bit rate 200,

bit rate 200 is divided by 4 to generate bit rate 50,

bit rate 2400 is divided by 18 to generate bit rate 134.5 with a frequency error of −0.87%,

bit rate 2400 is also divided by 22 to generate bit rate 110 with a frequency error of −0.83%, and

bit rate 9600 is divided by 16/3 to generate bit rate 1800.

The 16/3 division is accomplished by alternating the divide ratio between 5 (twice) and 6 (once). The result is an exact average output frequency with some frequency modulation. Taking advantage of the ÷ 16 feature of the UART, the resulting distortion is less than 0.78% regardless of the number of elements in a character, and therefore well within the timing accuracy specified for high speed communications equipment. All signals except 1800, have a 50% duty cycle.

Output Multiplexer — The outputs of the counter network are fed to a 16-input multiplexer, which is controlled by the Rate Select inputs (S_0-S_3). The multiplexer output is then resynchronized with the incoming clock in order to cancel all cumulative delays and to present an output signal at the buffered output (Z) that is synchronous with the prescaler outputs (Q_0-Q_2). Table 2 lists the correspondance between select code and output bit rate. Two of the 16 codes do not select an internally generated frequency, but select an input into which the user can feed either a different, non-standardized frequency, or a static level (HIGH or LOW) to generate "zero baud".

The bit rates most commonly used in modern data terminals (110, 150, 300, 1200, 2400 baud) require that no more than one input be grounded, easily achieved with a single pole, 5-position switch. 2400 baud is selected by two different codes, so that the whole spectrum of modern digital communication rates has a common HIGH on the S_3 input.

Initialization (Reset) — The initialization circuit generates a common master reset signal for all flip-flops in the IM4702/12. This signal is derived from a digital differentiator that senses the first HIGH level on the CP input after the $\overline{E}_{CP}$ input goes LOW. Upon initialization, all counters are reset and all outputs will be in the LOW state. When $\overline{E}_{CP}$ is HIGH, selecting the Crystal input, CP must be LOW; a HIGH level on CP would apply a continuous reset.

All inputs to the 4702/12 except I_X have on-chip pull-up circuits; the I_X input of the 4712 has a high value resistor tied to O_X.

Table 2: Truth Table for Rate Select Inputs

S_3	S_2	S_1	S_0	Output Rate (Z) Note 1
L	L	L	L	Multiplexed Input (I_M)
L	L	L	H	Multiplexed Input (I_M)
L	L	H	L	50 Baud
L	L	H	H	75 Baud
L	H	L	L	134.5 Baud
L	H	L	H	200 Baud
L	H	H	L	600 Baud
L	H	H	H	2400 Baud
H	L	L	L	9600 Baud
H	L	L	H	4800 Baud
H	L	H	L	1800 Baud
H	L	H	H	1200 Baud
H	H	L	L	2400 Baud
H	H	L	H	300 Baud
H	H	H	L	150 Baud
H	H	H	H	110 Baud

L = LOW Level
H = HIGH Level

Note 1: Actual output frequency is 16 times the indicated output rate, assuming a clock frequency of 2.4576MHz.

Table 3: Crystal Specifications

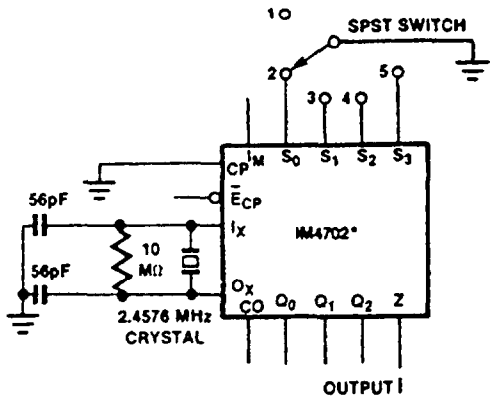
Parameters	Typical Crystal Spec
Frequency	2.4576MHz "AT" Cut
Series Resistance (Max)	250Ω
Unwanted Modes	−6dB (Min)
Type of Operation	Parallel
Load Capacitance	32pF ± 0.5pF

APPLICATIONS

Single Channel Bit Rate Generator

Figure 4 shows the simplest application of the IM4702/12. This circuit generates one of five possible bit rates as determined by the setting of a single pole, 5-position switch. The Bit Rate Output (Z) drives one standard TTL load or four low power Schottky loads over the full temperature range. The possible output frequencies correspond to 100, 150, 300, 1200, and 2400 or 3600 Baud. For many low cost terminals, these five bit rates are adequate.

This mode of operation is commonly chosen for applications using industry standard 1402/6402 UARTs.



Switch Position	Bit Rate
1	110 Baud
2	150 Baud
3	300 Baud
4	1200 Baud
5	2400 Baud

Figure 4: Switch Selectable Bit Rate Generator Configuration Providing Five Bit Rates

Simultaneous Generation of Several Bit Rates

Figure 5 shows a simple scheme that generates eight bit rates on eight output lines, using one IM4702/12 and one 93L34 Bit Addressable Latch. This and the following applications take advantage of the built-in scan counter (prescaler) outputs. As shown in the block diagram, these outputs (Q_0 to Q_2) go through a complete sequence of eight states for every half-period of the highest output frequency (9600 Baud). Feeding these Scan Counter Outputs back to the Select inputs of the multiplexer causes the IM4702/12 to sequentially interrogate the state of eight different frequency signals. The 93L34 Bit Addressable Latch, addressed by the same Scan Counter Outputs, reconverts the multiplexed single Output (Z) into eight parallel output frequency signals. In the simple scheme of Figure 5, input S_3 is left open (HIGH) and the following bit rates are generated:

Q_0: 110 Baud Q_3: 1800 Baud Q_6: 300 Baud
Q_1: 9600 Baud Q_4: 1200 Baud Q_7: 150 Baud
Q_2: 4800 Baud Q_5: 2400 Baud

Other bit rate combinations can be generated by changing the Scan Counter to Selector interconnection or by inserting logic gates into this path.

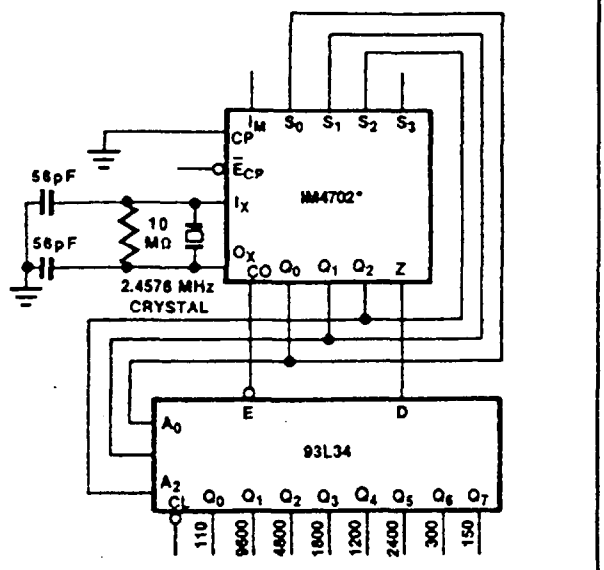


Figure 5: Bit Rate Generator Configuration With Eight Simultaneous Frequencies

19200 Baud Operation

A 19200 baud signal is available on the Q_2 output, but is not internally connected to the multiplexer. This signal can be generated on the Z output by connecting the Q_2 output to the I_M input and applying select code. An additional 2-input NOR gate can be used to retain the "Zero Baud" feature on select code 1 for the IM4702/12. (See Figure 6).

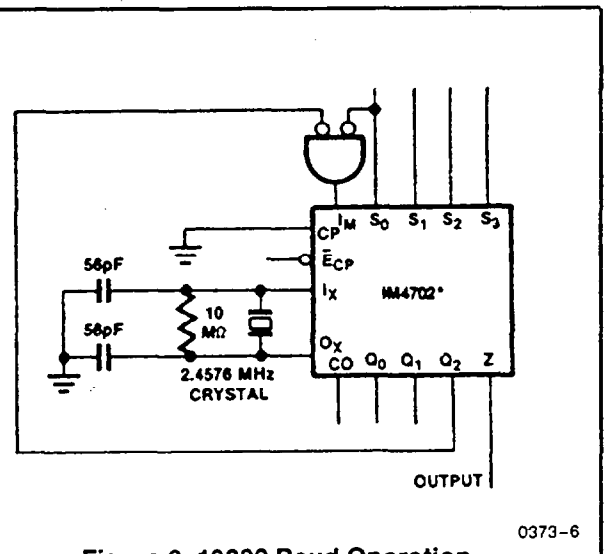


Figure 6: 19200 Baud Operation

* The 4712 may replace the 4702 in the above applications with the standard 2.4576MHz crystal. The two external capacitors and one resistor are not required when using the 4712.

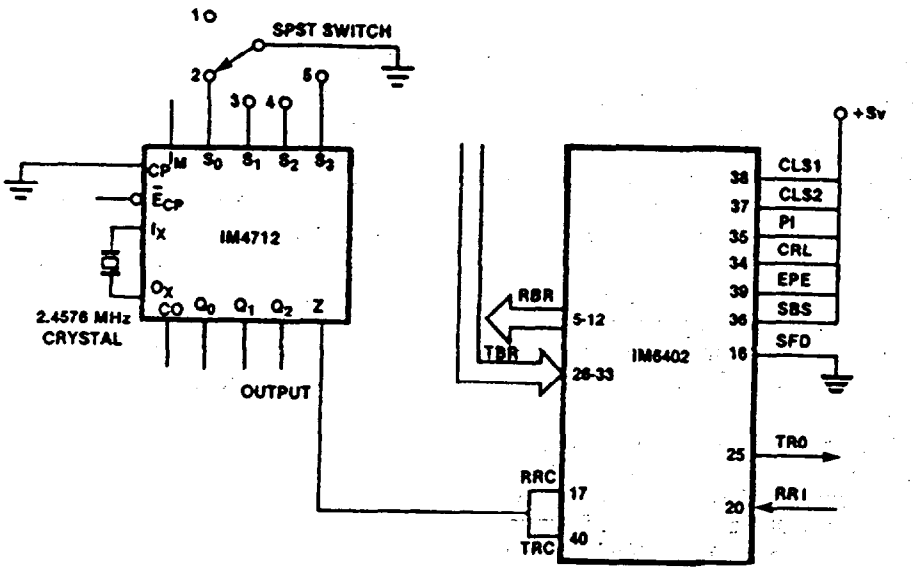


Figure 7: IM4712 Baud Rate Generator With IM6402 CMOS UART

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19-0175; Rev 3; 5/96

MAXIM

±15kV ESD-Protected, +5V RS-232 Transceivers

General Description

The MAX202E–MAX213E, MAX232E/MAX241E line drivers/receivers are designed for RS-232 and V.28 communications in harsh environments. Each transmitter output and receiver input is protected against ±15kV electrostatic discharge (ESD) shocks, without latchup. The various combinations of features are outlined in the *Selection Guide*. The drivers and receivers for all ten devices meet all EIA/TIA-232E and CCITT V.28 specifications at data rates up to 120kbps, when loaded in accordance with the EIA/TIA-232E specification.

The MAX211E/MAX213E/MAX241E are available in 28-pin SO packages, as well as a 28-pin SSOP that uses 60% less board space. The MAX202E/MAX232E come in 16-pin narrow SO, wide SO, and DIP packages. The MAX203E comes in a 20-pin DIP/SO package, and needs no external charge-pump capacitors. The MAX205E comes in a 24-pin wide DIP package, and also eliminates external charge-pump capacitors. The MAX206E/MAX207E/MAX208E come in 24-pin SO, SSOP, and narrow DIP packages. The MAX232E/MAX241E operate with four 1µF capacitors, while the MAX202E/MAX206E/MAX207E/MAX208E/MAX211E/MAX213E operate with four 0.1µF capacitors, further reducing cost and board space.

Applications

Notebook, Subnotebook, and Palmtop Computers
Battery-Powered Equipment
Hand-Held Equipment

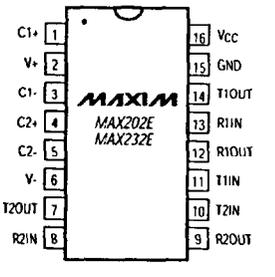
Ordering Information appears at end of data sheet.

Features

- **ESD Protection for RS-232 I/O Pins:**
  ±15kV—Human Body Model
  ±8kV—IEC1000-4-2, Contact Discharge
  ±15kV—IEC1000-4-2, Air-Gap Discharge
- **Latchup Free (unlike bipolar equivalents)**
- **Guaranteed 120kbps Data Rate—LapLink™ Compatible**
- **Guaranteed 3V/µs Min Slew Rate**
- **Operate from a Single +5V Power Supply**

Pin Configurations

TOP VIEW



DIP/SO

Pin Configurations and Typical Operating Circuits continued at end of data sheet.

Selection Guide

PART	No. of RS-232 DRIVERS	No. of RS-232 RECEIVERS	RECEIVERS ACTIVE IN SHUTDOWN	No. of EXTERNAL CAPACITORS	LOW-POWER SHUTDOWN	TTL THREE-STATE
MAX202E	2	2	0	4 (0.1µF)	No	No
MAX203E	2	2	0	None	No	No
MAX205E	5	5	0	None	Yes	Yes
MAX206E	4	3	0	4 (0.1µF)	Yes	Yes
MAX207E	5	3	0	4 (0.1µF)	No	No
MAX208E	4	4	0	4 (0.1µF)	No	No
MAX211E	4	5	0	4 (0.1µF)	Yes	Yes
MAX213E	4	5	2	4 (0.1µF)	Yes	Yes
MAX232E	2	2	0	4 (1µF)	No	No
MAX241E	4	5	0	4 (1µF)	Yes	Yes

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ABSOLUTE MAXIMUM RATINGS

VCC ..... -0.3V to +6V
V+ ..... (VCC - 0.3V) to +14V
V- ..... -14V to +0.3V
Input Voltages
T_IN ..... -0.3V to (V+ + 0.3V)
R_IN ..... ±30V
Output Voltages
T_OUT ..... (V- - 0.3V) to (V+ + 0.3V)
R_OUT ..... -0.3V to (VCC + 0.3V)
Short-Circuit Duration, T_OUT ..... Continuous
Continuous Power Dissipation (T_A = +70°C)
16-Pin Plastic DIP (derate 10.53mW/°C above +70°C) 842mW
16-Pin Narrow SO (derate 8.70mW/°C above +70°C) 696mW
16-Pin Wide SO (derate 9.52mW/°C above +70°C) 762mW
20-Pin Plastic DIP (derate 11.11mW/°C above +70°C) 889mW
20-Pin SO (derate 10.00mW/°C above +70°C) 800mW
24-Pin Narrow Plastic DIP
(derate 13.33mW/°C above +70°C) 1.07W
24-Pin Wide Plastic DIP
(derate 14.29mW/°C above +70°C) 1.14W
24-Pin SO (derate 11.76mW/°C above +70°C) 941mW
24-Pin SSOP (derate 8.00mW/°C above +70°C) 640mW
28-Pin SO (derate 12.50mW/°C above +70°C) 1W
28-Pin SSOP (derate 9.52mW/°C above +70°C) 762mW
Operating Temperature Ranges
MAX2_ _EC_ _ ..... 0°C to +70°C
MAX2_ _EE_ _ ..... -40°C to +85°C
Storage Temperature Range ..... -65°C to +165°C
Lead Temperature (soldering, 10sec) ..... +300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(VCC = +5V ±10% for MAX202E/206E/208E/211E/213E/232E/241E; VCC = +5V ±5% for MAX203E/205E/207E; C1–C4 = 0.1µF for MAX202E/206E/207E/208E/211E/213E; C1–C4 = 1µF for MAX232E/241E; T_A = T_{MIN} to T_{MAX}; unless otherwise noted. Typical values are at T_A = +25°C.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
DC CHARACTERISTICS							
VCC Supply Current	ICC	No load, T_A = +25°C	MAX202E/203E		8	15	mA
			MAX205E-208E		11	20	
			MAX211E/213E		14	20	
			MAX232E		5	10	
			MAX241E		7	15	
Shutdown Supply Current		T_A = +25°C, Figure 1	MAX205E/206E		1	10	µA
			MAX211E/241E		1	10	
			MAX213E		15	50	
LOGIC							
Input Pull-Up Current		T_IN = 0V (MAX205E-208E/211E/213E/241E)			15	200	µA
Input Leakage Current		T_IN = 0V to VCC (MAX202E/203E/232E)				±10	µA
Input Threshold Low	VIL	T_IN; EN, $\overline{SHDN}$ (MAX213E) or $\overline{EN}$, SHDN (MAX205E-208E/211E/241E)				0.8	V
Input Threshold High	VIH	T_IN		2.0			V
		EN, $\overline{SHDN}$ (MAX213E) or $\overline{EN}$, SHDN (MAX205E-208E/211E/241E)		2.4			
Output Voltage Low	VOL	R_OUT; IOUT = 3.2mA (MAX202E/203E/232E) or IOUT = 1.6mA (MAX205E/208E/211E/213E/241E)				0.4	V
Output Voltage High	VOH	R_OUT; IOUT = -1.0mA		3.5	VCC - 0.4		V
Output Leakage Current		$\overline{EN}$ = VCC, EN = 0V, 0V ≤ ROUT ≤ VCC, MAX205E-208E/211E/213E/241E outputs disabled			±0.05	±10	µA

ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = +5V ±10% for MAX202E/206E/208E/211E/213E/232E/241E; V_{CC} = +5V ±5% for MAX203E/205E/207E; C1–C4 = 0.1µF for MAX202E/206E/207E/208E/211E/213E; C1–C4 = 1µF for MAX232E/241E; T_A = T_{MIN} to T_{MAX}; unless otherwise noted. Typical values are at T_A = +25°C.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
EIA/TIA-232E RECEIVER INPUTS							
Input Voltage Range				-30		30	V
Input Threshold Low		T_A = +25°C, V_{CC} = 5V	All parts, normal operation	0.8	1.2		V
			MAX213E, $\overline{SHDN}$ = 0V, EN = V_{CC}	0.6	1.5		
Input Threshold High		T_A = +25°C, V_{CC} = 5V	All parts, normal operation		1.7	2.4	V
			MAX213E (R4, R5), $\overline{SHDN}$ = 0V, EN = V_{CC}		1.5	2.4	
Input Hysteresis		V_{CC} = 5V, no hysteresis in shutdown		0.2	0.5	1.0	V
Input Resistance		T_A = +25°C, V_{CC} = 5V		3	5	7	kΩ
EIA/TIA-232E TRANSMITTER OUTPUTS							
Output Voltage Swing		All drivers loaded with 3kΩ to ground (Note 1)		±5	±9		V
Output Resistance		V_{CC} = V+ = V- = 0V, V_{OUT} = ±2V		300			Ω
Output Short-Circuit Current					±10	±60	mA
TIMING CHARACTERISTICS							
Maximum Data Rate		R_L = 3kΩ to 7kΩ, C_L = 50pF to 1000pF, one transmitter switching		120			kbps
Receiver Propagation Delay	t_{PLHR}, t_{PHLR}	C_L = 150pF	All parts, normal operation		0.5	10	µs
			MAX213E (R4, R5), $\overline{SHDN}$ = 0V, EN = V_{CC}		4	40	
Receiver Output Enable Time		MAX205E/206E/211E/213E/241E normal operation, Figure 2			600		ns
Receiver Output Disable Time		MAX205E/206E/211E/213E/241E normal operation, Figure 2			200		ns
Transmitter Propagation Delay	t_{PLHT}, t_{PHLT}	R_L = 3kΩ, C_L = 2500pF, all transmitters loaded			2		µs
Transition-Region Slew Rate		T_A = +25°C, V_{CC} = 5V, R_L = 3kΩ to 7kΩ, C_L = 50pF to 1000pF, measured from -3V to +3V or +3V to -3V, Figure 3		3	6	30	V/µs
ESD PERFORMANCE: TRANSMITTER OUTPUTS, RECEIVER INPUTS							
ESD-Protection Voltage		Human Body Model			±15		kV
		IEC1000-4-2, Contact Discharge			±8		
		IEC1000-4-2, Air-Gap Discharge			±15		

Note 1: MAX211EE_ _ tested with V_{CC} = +5V ±5%.

Pin Descriptions

MAX202E/MAX232E

PIN		NAME	FUNCTION
DIP/SO	LCC		
1, 3	2, 4	C1+, C1-	Terminals for positive charge-pump capacitor
2	3	V+	+2V_{CC} voltage generated by the charge pump
4, 5	5, 7	C2+, C2-	Terminals for negative charge-pump capacitor
6	8	V-	-2V_{CC} voltage generated by the charge pump
7, 14	9, 18	T_OUT	RS-232 Driver Outputs
8, 13	10, 17	R_IN	RS-232 Receiver Outputs
9, 12	12, 15	R_OUT	RS-232 Receiver Outputs
10, 11	13, 14	T_IN	RS-232 Driver Inputs
15	19	GND	Ground
16	20	V_{CC}	+4.5V to +5.5V Supply-Voltage Input
—	1, 6, 11, 16	N.C.	No Connect—not internally connected.

MAX203E

PIN		NAME	FUNCTION
DIP	SO		
1, 2	1, 2	T_IN	RS-232 Driver Inputs
3, 20	3, 20	R_OUT	RS-232 Receiver Outputs
4,19	4, 19	R_IN	RS-232 Receiver Inputs
5,18	5, 18	T_OUT	RS-232 Transmitter Outputs
6, 9	6, 9	GND	Ground
7	7	V_{CC}	+4.5V to +5.5V Supply-Voltage Input
8	13	C1+	Make no connection to this pin.
10, 16	11, 16	C2-	Connect pins together.
12, 17	10, 17	V-	-2V_{CC} voltage generated by the charge pump. Connect pins together.
13	14	C1-	Make no connection to this pin.
14	8	V+	+2V_{CC} voltage generated by the charge pump
11, 15	12, 15	C2+	Connect pins together.

MAX205E

PIN	NAME	FUNCTION
1–4, 19	T_OUT	RS-232 Driver Outputs
5, 10, 13, 18, 24	R_IN	RS-232 Receiver Inputs
6, 9, 14, 17, 23	R_OUT	TTL/CMOS Receiver Outputs. All receivers are inactive in shutdown.
7, 8, 15, 16, 22	T_IN	TTL/CMOS Driver Inputs. Internal pull-ups to V_{CC}.
11	GND	Ground
12	V_{CC}	+4.75V to +5.25V Supply Voltage
20	$\overline{EN}$	Receiver Enable—active low
21	SHDN	Shutdown Control—active high

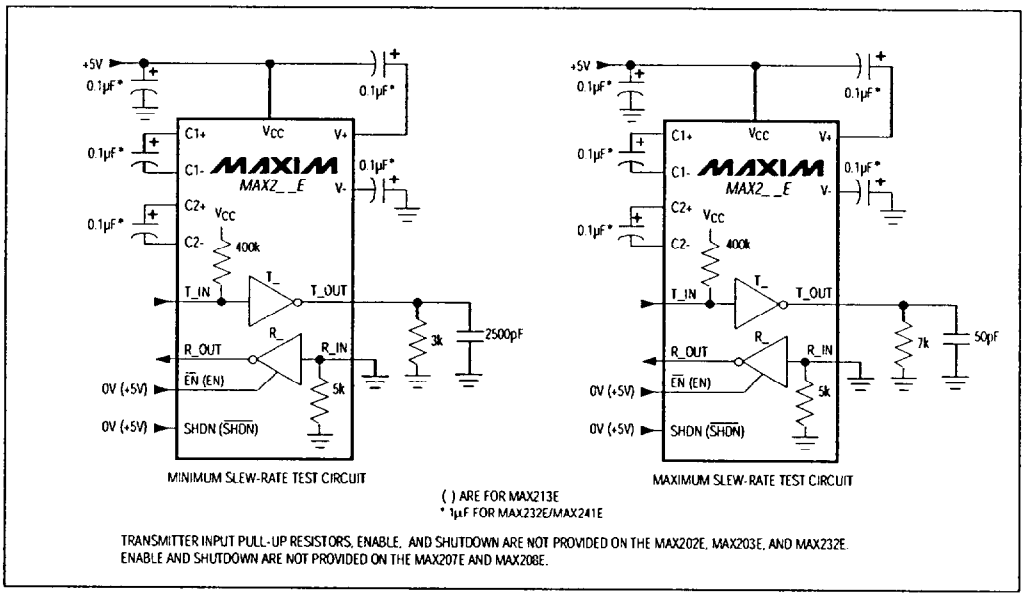


Figure 3. Transition Slew-Rate Circuit

Detailed Description

The MAX202E–MAX213E, MAX232E/MAX241E consist of three sections: charge-pump voltage converters, drivers (transmitters), and receivers. These E versions provide extra protection against ESD. They survive ±15kV discharges to the RS-232 inputs and outputs, tested using the Human Body Model. When tested according to IEC1000-4-2, they survive ±8kV contact-discharges and ±15kV air-gap discharges. The rugged E versions are intended for use in harsh environments or applications where the RS-232 connection is frequently changed (such as notebook computers). The standard (non-"E") MAX202, MAX203, MAX205–MAX208, MAX211, MAX213, MAX232, and MAX241 are recommended for applications where cost is critical.

+5V to ±10V Dual Charge-Pump Voltage Converter

The +5V to ±10V conversion is performed by dual charge-pump voltage converters (Figure 4). The first charge-pump converter uses capacitor C1 to double the +5V into +10V, storing the +10V on the output filter capacitor, C3. The second uses C2 to invert the +10V into -10V, storing the -10V on the V- output filter capacitor, C4.

In shutdown mode, V+ is internally connected to VCC by a 1kΩ pull-down resistor, and V- is internally connected to ground by a 1kΩ pull-up resistor.

RS-232 Drivers

With VCC = 5V, the typical driver output voltage swing is ±8V when loaded with a nominal 5kΩ RS-232 receiver. The output swing is guaranteed to meet EIA/TIA-232E and V.28 specifications that call for ±5V minimum output levels under worst-case conditions. These include a 3kΩ load, minimum VCC, and maximum operating temperature. The open-circuit output voltage swings from (V+ - 0.6V) to V-.

Input thresholds are CMOS/TTL compatible. The unused drivers' inputs on the MAX205E–MAX208E, MAX211E, MAX213E, and MAX241E can be left unconnected because 400kΩ pull-up resistors to VCC are included on-chip. Since all drivers invert, the pull-up resistors force the unused drivers' outputs low. The MAX202E, MAX203E, and MAX232E do not have pull-up resistors on the transmitter inputs.

When in low-power shutdown mode, the MAX205E/MAX206E/MAX211E/MAX213E/MAX241E driver outputs are turned off and draw only leakage currents—even if they are back-driven with voltages between 0V and 12V. Below -0.5V in shutdown, the transmitter output is diode-clamped to ground with a 1kΩ series impedance.

RS-232 Receivers

The receivers convert the RS-232 signals to CMOS-logic output levels. The guaranteed 0.8V and 2.4V receiver input thresholds are significantly tighter than the ±3V thresholds required by the EIA/TIA-232E specification. This allows the receiver inputs to respond to TTL/CMOS-logic levels, as well as RS-232 levels.

The guaranteed 0.8V input low threshold ensures that receivers shorted to ground have a logic 1 output. The 5kΩ input resistance to ground ensures that a receiver with its input left open will also have a logic 1 output.

Receiver inputs have approximately 0.5V hysteresis. This provides clean output transitions, even with slow rise/fall-time signals with moderate amounts of noise and ringing.

In shutdown, the MAX213E's R4 and R5 receivers have no hysteresis.

Shutdown and Enable Control (MAX205E/MAX206E/MAX211E/MAX213E/MAX241E)

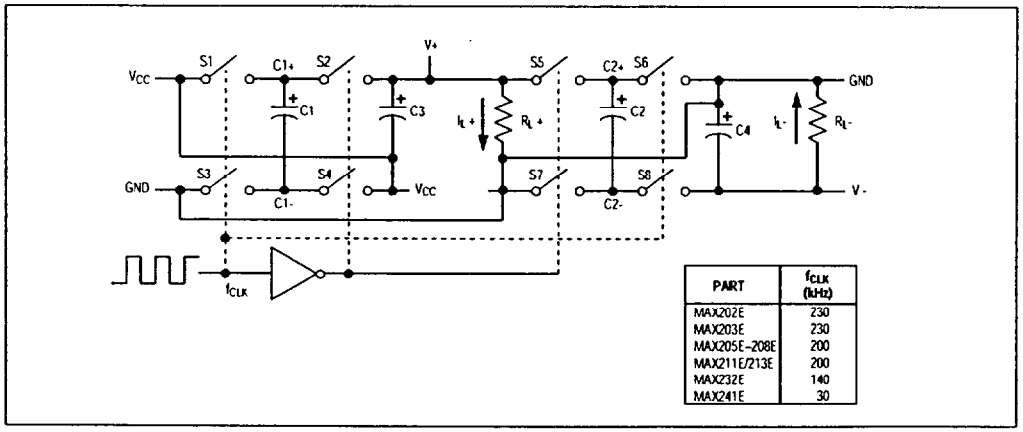
In shutdown mode, the charge pumps are turned off, V+ is pulled down to VCC, V- is pulled to ground, and the transmitter outputs are disabled. This reduces supply current typically to 1µA (15µA for the MAX213E). The time required to exit shutdown is under 1ms, as shown in Figure 5.

Receivers

All MAX213E receivers, except R4 and R5, are put into a high-impedance state in shutdown mode (see Tables 1a and 1b). The MAX213E's R4 and R5 receivers still function in shutdown mode. These two awake-in-shutdown receivers can monitor external activity while maintaining minimal power consumption.

The enable control is used to put the receiver outputs into a high-impedance state, to allow wire-OR connection of two EIA/TIA-232E ports (or ports of different types) at the UART. It has no effect on the RS-232 drivers or the charge pumps.

Note: The enable control pin is active low for the MAX211E/MAX241E ($\overline{EN}$), but is active high for the MAX213E (EN). The shutdown control pin is active high for the MAX205E/MAX206E/MAX211E/MAX241E (SHDN), but is active low for the MAX213E ($\overline{SHDN}$).



PART	f_{CLK} (kHz)
MAX202E	230
MAX203E	230
MAX205E–208E	200
MAX211E/213E	200
MAX232E	140
MAX241E	30

Figure 4. Charge-Pump Diagram

Applications Information

Capacitor Selection

The capacitor type used for C1–C4 is not critical for proper operation. The MAX202E, MAX206–MAX208E, MAX211E, and MAX213E require 0.1µF capacitors, and the MAX232E and MAX241E require 1µF capacitors, although in all cases capacitors up to 10µF can be used without harm. Ceramic, aluminum-electrolytic, or tantalum capacitors are suggested for the 1µF capacitors, and ceramic dielectrics are suggested for the 0.1µF capacitors. When using the minimum recommended capacitor values, make sure the capacitance value does not degrade excessively as the operating temperature varies. If in doubt, use capacitors with a larger (e.g., 2x) nominal value. The capacitors' effective series resistance (ESR), which usually rises at low temperatures, influences the amount of ripple on V+ and V-.

Use larger capacitors (up to 10µF) to reduce the output impedance at V+ and V-. This can be useful when "stealing" power from V+ or from V-. The MAX203E and MAX205E have internal charge-pump capacitors.

Bypass V_{CC} to ground with at least 0.1µF. In applications sensitive to power-supply noise generated by the charge pumps, decouple V_{CC} to ground with a capacitor the same size as (or larger than) the charge-pump capacitors (C1–C4).

V+ and V- as Power Supplies

A small amount of power can be drawn from V+ and V-, although this will reduce both driver output swing and noise margins. Increasing the value of the charge-pump capacitors (up to 10µF) helps maintain performance when power is drawn from V+ or V-.

Driving Multiple Receivers

Each transmitter is designed to drive a single receiver. Transmitters can be paralleled to drive multiple receivers.

Driver Outputs when Exiting Shutdown

The driver outputs display no ringing or undesirable transients as they come out of shutdown.

High Data Rates

These transceivers maintain the RS-232 ±5.0V minimum driver output voltages at data rates of over 120kbps. For data rates above 120kbps, refer to the Transmitter Output Voltage vs. Load Capacitance graphs in the *Typical Operating Characteristics*. Communication at these high rates is easier if the capacitive loads on the transmitters are small; i.e., short cables are best.

Table 2. Summary of EIA/TIA-232E, V.28 Specifications

PARAMETER		CONDITIONS	EIA/TIA-232E, V.28 SPECIFICATIONS
Driver Output Voltage	0 Level	3kΩ to 7kΩ load	+5V to +15V
	1 Level	3kΩ to 7kΩ load	-5V to -15V
Driver Output Level, Max		No load	±25V
Data Rate		$3k\Omega \le R_L \le 7k\Omega$, $C_L \le 2500pF$	Up to 20kbps
Receiver Input Voltage	0 Level		+3V to +15V
	1 Level		-3V to -15V
Receiver Input Level			±25V
Instantaneous Slew Rate, Max		$3k\Omega \le R_L \le 7k\Omega$, $C_L \le 2500pF$	30V/µs
Driver Output Short-Circuit Current, Max			100mA
Transition Rate on Driver Output		V.28	1ms or 3% of the period
		EIA/TIA-232E	4% of the period
Driver Output Resistance		$-2V < V_{OUT} < +2V$	300Ω

Table 3. DB9 Cable Connections Commonly Used for EIA/TIAE-232E and V.24 Asynchronous Interfaces

PIN	CONNECTION	
1	Received Line Signal Detector (sometimes called Carrier Detect, DCD)	Handshake from DCE
2	Receive Data (RD)	Data from DCE
3	Transmit Data (TD)	Data from DTE
4	Data Terminal Ready	Handshake from DTE
5	Signal Ground	Reference point for signals
6	Data Set Ready (DSR)	Handshake from DCE
7	Request to Send (RTS)	Handshake from DTE
8	Clear to Send (CTS)	Handshake from DCE
9	Ring Indicator	Handshake from DCE

Pin Configurations and Typical Operating Circuits (continued)